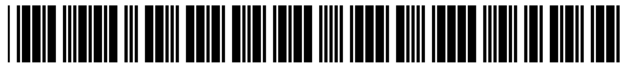


United States Patent [19]

Guo

[11] **Patent Number: 5,837,573**

[45] **Date of Patent: Nov. 17, 1998**

[54] **MINIMUM SIZE INTEGRATED CIRCUIT STATIC MEMORY CELL**

[75] Inventor: **Jeng-Jong Guo**, Hsin-Chu, Taiwan

[73] Assignee: **Utron Technology Inc.**, Hsin-Chu, Taiwan

[21] Appl. No.: **929,305**

[22] Filed: **Sep. 10, 1997**

Related U.S. Application Data

[62] Division of Ser. No. 460,035, Jun. 2, 1995, Pat. No. 5,703,392.

[51] **Int. Cl.**[6] **H01L 21/8244**

[52] **U.S. Cl.** **438/238**; 438/241; 438/155

[58] **Field of Search** 438/155, 238, 438/241, 212, 220, 429

[56] **References Cited**

U.S. PATENT DOCUMENTS

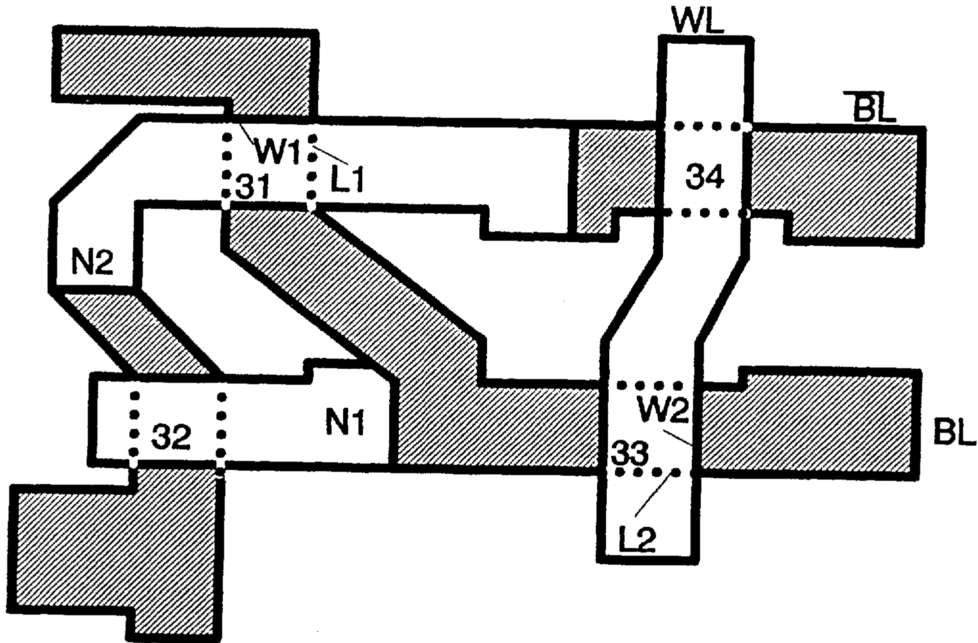
5,703,392 12/1997 Guo 257/392

Primary Examiner—Peter Toby Brown
Assistant Examiner—Toniae M. Thomas
Attorney, Agent, or Firm—H. C. Lin

[57] **ABSTRACT**

A semiconductor static memory cell with two cross-coupled inverters and two transmission gates for coupling two bit lines uses all minimum size (gate length and gate width) MOSFETs to achieve minimum area. This minimum dimension is rendered possible by using a higher threshold voltage for the transmission gate MOSFET than the threshold voltage of pull-down MOSFET of the inverter. Different threshold voltages are obtained with selective ion implantation, different gate oxide thicknesses and/or different gate doping.

6 Claims, 2 Drawing Sheets



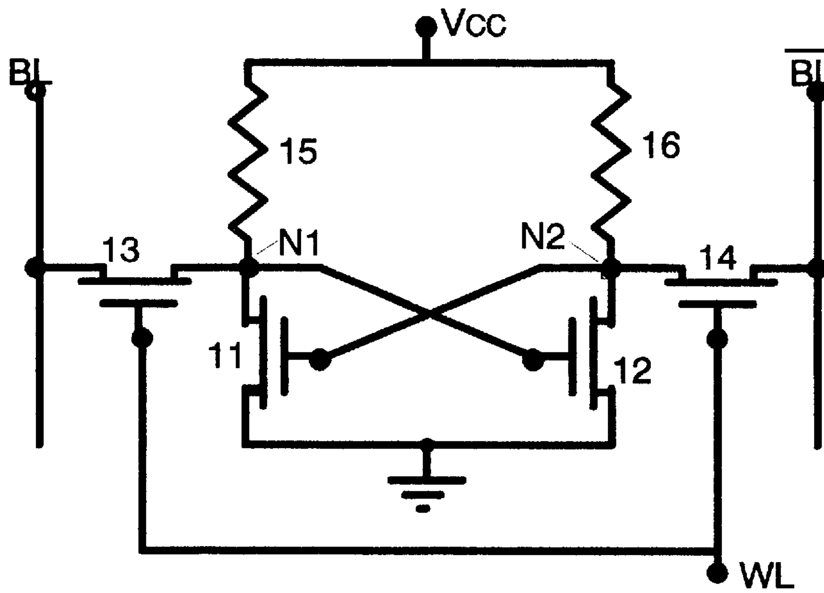


FIG. 1 (PRIOR ART)

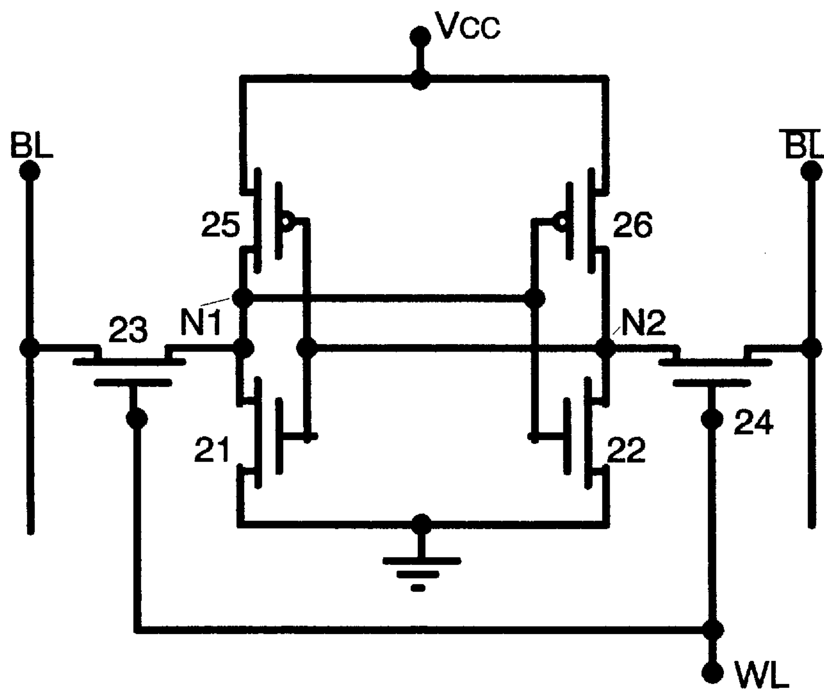


FIG. 2 (PRIOR ART)

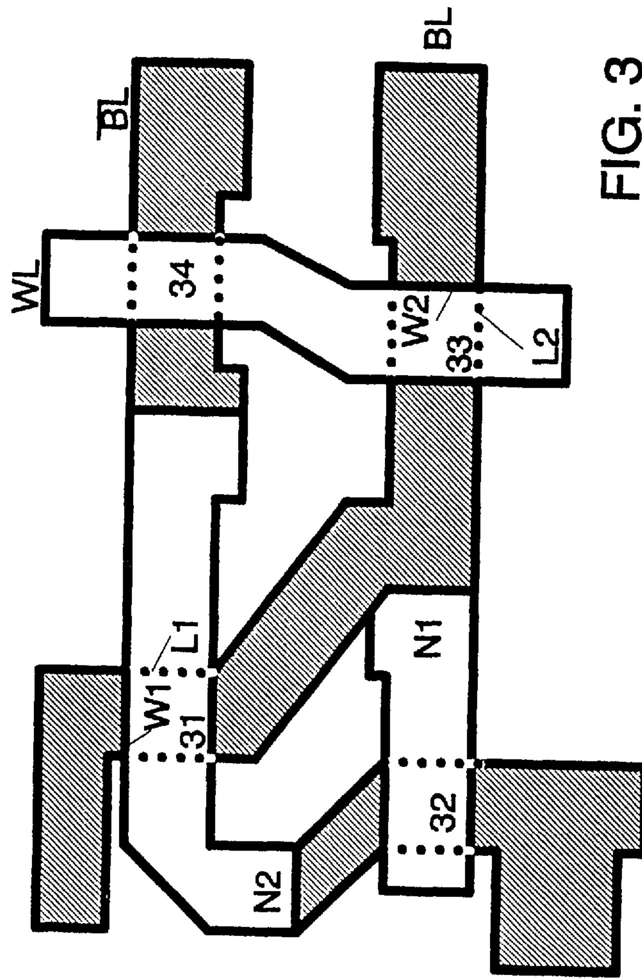


FIG. 3

MINIMUM SIZE INTEGRATED CIRCUIT STATIC MEMORY CELL

This application is a division of application Ser. No. 08/460,035, filed Jun. 2, 1995, U.S. Pat. No. 5,703,392.

INTRODUCTION

This invention is related to integrated circuit memory cells, in particular to static memory cells.

Semicondcutor memory is one of the most widely used integrated circuits today. Semiconductor memories can be subdivided into random-access memory (RAM) and read-only memory (ROM). There are two major types of RAMs, namely: the static memory (SRAM) and the dynamic memory (DRAM). The dynamic memory stores binary data in a capacitance, but needs refreshing periodically to replenish the charge stored in the capacitance. The SRAM does not require refreshing in its operation, but occupies a larger area, because the SRAM cell has more transistors than a DRAM cell. It is desirable to minimize the area of an SRAM cell to achieve high package density.

A SRAM cell generally includes two cross-coupled inverters to form a flip-flop or latch. The inverter can use a polycrystalline silicon thin film load resistance as shown in FIG. **1**. The inverter can also be a complementary MOSFET (CMOS) inverter as shown in FIG. **2**. In either circuit, low power dissipation can be achieved. In the case of the polycrystalline semiconductor thin film load resistance, a very high value load resistance is used to minimize the load current, utilizing the high sheet resistivity of the polycrystalline silicon thin film. In the case of the CMOS inverter, the pull-up p-channel MOSFET (PMOS) and the pull-down n-channel MOSFET (NMOS) are never turned on at the same time in the steady state condition and therefore the series current is nearly equal to zero.

In FIG. **1**, an NMOS **11** with its source connected to ground is pulled up by a load resistor **15** toward a positive power supply VDD to constitute a first inverter. Another NMOS **12** is similarly pulled up by resistor **16** to constitute a second inverter. These two inverters are cross-coupled to each other, such that the node N**1** of the first inverter is connected to the gate of the NMOS **12**, and the node N**2** of the second inverter is connected to the gate of the NMOS **11**. The node N**1** is connected to a bit line BL through a pass transistor NMOS **13** as a transmission gate, and the node N**2** is connected to the complementary bit line BL through another pass transistor NMOS **14**. The gates of the pass transistors **13** and **14** are connected to a word line WL.

Similarly, the CMOS memory cell shown in FIG. **2** has two CMOS inverters. The first inverter has a PMOS **25** connected in series with an NMOS **21** and a common gate connection. The second inverter has a PMOS **26** connected in series with an NMOS **22** and a common gate connection. The node N**1** at the common drain of PMOS **25** and NMOS **21** is connected to the common gate of PMOS **26** and NMOS **22**. The node at the common drain of PMOS **26** and NMOS **22** is connected to the common gate of PMOS **25** and NMOS **21**. The node N**1** is connected to a bit line BL through a pass transistor **23**, and the node N**2** is connected to a complementary bit line $\overline{BL}$ through another pass transistor **24**. The gates of the pass transistors **23** and **24** are connected to a word line WL.

For minimum cell area, it is desirable to make all the MOSFETs of minimum size. However, due to noise margin considerations, it is not feasible to make all the transistors of minimum size. For reliable noise margin, the "0" logic level appearing at the node N**1** and N**2** should be low enough. For a lower voltage drop, the conventional memory cell typically makes the width-to-length ratio of the gates of NMOS **11** or **21** 2.5 to 4 times larger than the width-to-length ratio of the NMOS transmission gates **13** or **23**. When two transistors have different width-to-length ratios, the two transistors cannot be both of minimum size.

In U.S. Pat. No. 5,047,979, a memory cell of similar channel width was disclosed. However, the channel length was not made equal in order to satisfy the requirement of larger width-to-length ratio of the pull-down NMOS than that of the transmission gate. Thus, a minimum size memory cell was not achieved.

SUMMARY

An object of this invention is to minimize the area of a static memory cell. Another object of this invention is to use minimum dimensions for all the transistors in a MOSFET static memory cell. Still another object of this invention is to use nearly equal width-to-length ratio gates for all the transistors in a MOSFET static memory cell.

These objects are achieved in this invention by using a higher threshold voltage for the transmission gate of the SRAM cell than the threshold voltage for the pull-down MOSFET of the flip-flop, which is coupled to the bit lines through the transmission gates. The higher threshold is achieved by doping the substrate, thickening the gate oxide or doping the gate with appropriate impurities.

These objects are further achieved in this invention by using a lower transconductance parameter KP for the pass transistor than that of the pull-down MOSFET through adjusting the thickness of respective gate oxides.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. **1** shows the circuit diagram of a static memory cell using thin film resistors as load devices.

FIG. **2** shows the circuit diagram of static memory cell using CMOS.

FIG. **3** shows the layout of a memory cell, based on the present invention in which minimum gate width and gate length are used.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENT

In this invention, a higher threshold voltage is used for the transmission gates **13**, **14** than that of the pull-down NMOS **11**, **12** in FIG. **1**; and a higher threshold voltage is used for the transmission gates **23**, **24** than that of the pull-down NMOS **21**, **22** in FIG. **2**. Gates of minimum length and minimum width or nearly equal width-to-length ratio are used for all the MOSFETs. As a limit, the width-to-length ratio (W**1**/L**1** in FIG. **3**) of the gates of the pull-down MOSFET is less than 2.5 times the width-to-length ratio (W**2**/L**2** in FIG. **3**) of the transmission gates. A possible layout is shown in FIG. **3** which corresponds to the circuit of FIG. **1**.

The drain to source voltage drop V_{DS} across a MOSFET conducting a current I_D is given as:

$$I_D \tilde{=} KP(W/L)(V_{GS}-Vth)V_{DS} \tag{1}$$

where KP is a transconductance parameter equal to the product of mobility u and specific gate capacitance C_{ox}, W is the gate width and L is the gate length. It can be seen that for the same W/L ratio, an increase in Vth of the transmis-

sion gate increases the ratio of the V_{DS} of the transmission gate to the V_{DS} of the pull-down NMOS, hence decreasing the node voltage **N1** or **N2**.

The threshold voltage of an NMOS is given by the following expression:

$$Vth=\phi_{GC}-2\phi_f-Q_{ss}/C_{ox}-Q_{BO}+\gamma[(-2\phi_f+V_{SB})^{1/2}-(-2\phi_f)^{1/2}] \quad (2)$$

where

ϕ_{GC} is the work function difference between the gate and the channel,

$2\phi_f$ is the surface potential,

Q_{ss} is the surface-state charge at the bulk-oxide interface,

Q_{BO} is charge density in the depletion layer of the channel at the onset of strong inversion,

γ is the body-effect coefficient and is equal to

$$(2q\epsilon_{si}N_a)^{1/2}/C_{ox},$$

q is the charge of an electron,

ϵ_{si} is the permittivity of silicon,

N_a is the background impurity concentration,

C_{ox} is the gate oxide specific capacitance and is equal to

$$\epsilon_{ox}/t_{ox},$$

ϵ_{ox} is the permittivity of gate oxide,

t_{ox} is the gate oxide thickness.

From this equation, it can be seen that the threshold voltage can be changed by adjusting several physical parameters in this equation.

In this invention, it is proposed to vary the work function difference, the background surface impurity concentration and/or the gate oxide thickness to obtain the desired threshold voltage.

The work function difference is the difference in Fermi levels of the silicon and gate materials, and can be hanged by doping the gate of the MOSFET with different conductivity type impurities. By changing the doping of a silicon gate from one heavily-doped conductivity-type (e.g. p-type) to a heavily-doped opposite conductivity-type (e.g. n-type), the work function difference can vary as much as one energy gas, i.e. 1.12V for silicon. For an NMOS, a p-type dopant such as boron decreases the work function difference between the gate and the n-type channel. Thus the threshold voltage of an NMOS with p-type silicon gate is higher than the threshold voltage of an NMOS with n-type silicon gate. Such a gate can be used to increase the threshold voltage of the NMOS pass transistors in the SRAM cell.

The threshold voltage can also be increased by using a thicker gate oxide due to decreased specific capacitance as shown in the above V_{th} equation 2.

A thicker oxide also decreases the transconductance parameter KP, which is equal to u^*C_{ox}, and thus effectively increases the threshold voltage, because a larger gate voltage is required to yield the same drain current as shown in Eq.1. Therefore, the objects of this invention can be implemented by using a lower transconductance parameter KP for the pass transistor than that of the pull-down MOSFET through adjusting the thickness of each gate oxide.

The threshold voltage can most readily adjusted by ion implantation. The change in threshold voltage is equal to Q_I/C_{ox}, where Q_I is the ion implantation dosage per unit area. When the implanted ions are of the same conductivity-type as the background, the threshold voltage is increased. When the implanted ions are of the opposite conductivity-type as the substrate, the threshold voltage is decreased. Furthermore, the ion-implantation also changes the body effect.

The above methods can be used either to change the threshold voltage of the transmission gates **13**, **14** in FIG. **1** or **23**, **24** in FIG. **2**. Conversely, the same principle can be used to decrease the threshold voltages of the pull-down transistors **11**, **12** in FIG. **1** or **21**, **22** in FIG. **2**. The choice is based on practical considerations. For instance, the gate oxide thickness cannot be reduced to a value where the integrity of the oxide becomes unreliable.

The above methods of varying the threshold voltage can be adopted by changing one or more parameters.

FIG. **3** shows a possible layout based on the present invention. The layout corresponds to the circuit diagram of FIG. **1**. In this layout, the areas **31** and **32** correspond to the pull-down MOSFETs **11** and **12** respectively. The areas **33** and **34** correspond to the pass transistors **13** and **14** respectively. Note that the gate length L, (shown as dotted lines), and the gate width W, (shown as the line sections bounded between the dotted lines) of all the transistors are all of equal minimum dimensions. While the figure is drawn with equal gate length and gate width, the W/L ratios need not be exactly equal to one. So long as the W/L ratio of the pull-down MOSFETs is approximately the same as the W/L ratio of the pass transistors within a factor of 2.5, this feature is within the scope of this invention.

While the foregoing descriptions are devoted to resistive loads and CMOS structures, the techniques apple equally to other types of memory cells. Memory cells such as those using a depletion-mode load, a pre-charged load and ferro-electric load are also within the scope of this invention.

What is claimed is:

1. A method to manufacture an integrates circuit static memory cell having two MOSFET pass transistors coupled to two cross-coupled inverters, comprising the step of adjusting the difference between the threshold voltage of pass transistor and the threshold voltage of the pull-down MOSFET of said inverters to be positive, wherein the width-to-length ratio of the gate of the pull-down MOSFET is less than 2.5 times the width-to-length ratio of the gate of the pass transistor.

2. A method to manufacture an integrated circuit static memory cell as described in **1**, wherein the threshold voltage of the MOSFET pass transistors is adjusted with a background surface dopant concentration of the substrate of said transistors, which is higher than and of same conductivity-type as the background surface dopant concentration of the pull-down MOSFET.

3. A method to manufacture an integrated circuit static memory cell as described in claim **1**, wherein the threshold voltage of the pull-down MOSFET is adjusted by reducing background surface dopant concentration of the substrate of said transistors.

4. A method to manufacture an integrated circuit static memory cell as described in claim **1**, wherein the threshold voltage of said pass transistors is adjusted with a gate oxide thicker than the gate oxide of the pull-down MOSFET.

5. A method to manufacture an integrated circuit static memory cell as described in claim **1**, where in the threshold voltage of the pull-down MOSFET is adjusted with a gate oxide thinner than the gate oxide of the pass transistors.

6. A method to manufacture an integrated circuit static memory cell as described in claim **1**, wherein the gate of the pass transistors is doped with accepter impurity arid the gate of the pull-down transistor is doped with donor impurity.

* * * * *